
Infinite-fold enhancement in communications capacity using pre-shared entanglement

Saikat Guha

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College of Optical Sciences,
Department of Electrical and Computer Engineering
University of Arizona, Tucson AZ (USA)



arXiv:2001.03934 (2020)

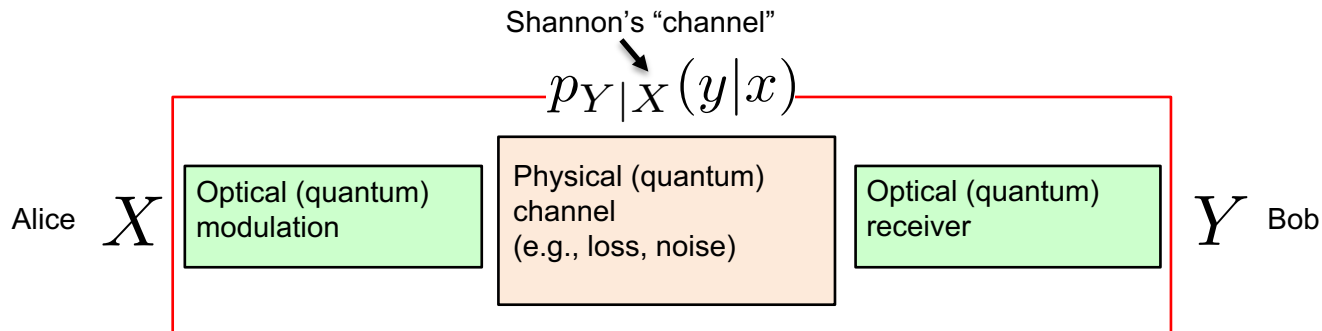
Outline

- Background
 - Quantum limit of classical communications
 - Joint detection receivers for superadditive capacity
- Entanglement assisted communications over the bosonic channel
 - Transmitter-receiver design
 - Infinite-fold capacity enhancement with pre-shared entanglement
- Conclusions and ongoing work

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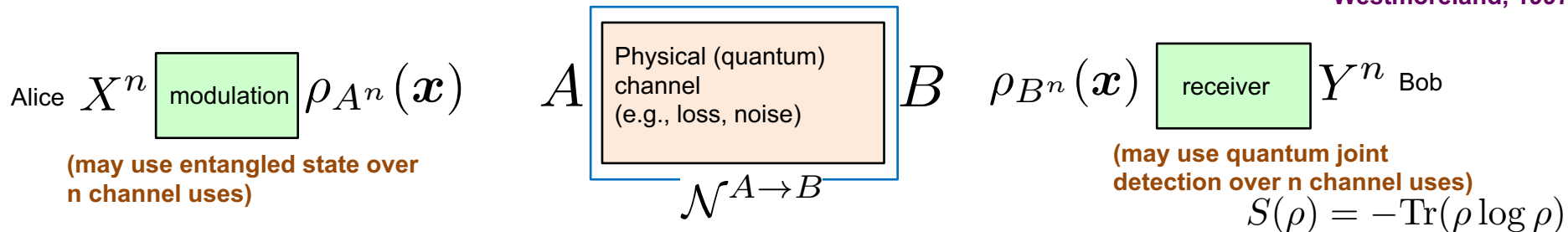
Quantum limit of classical communications



Shannon, 1948

Shannon capacity, $C_{\text{Shannon}} = \max_{p_X} I(X; Y)$

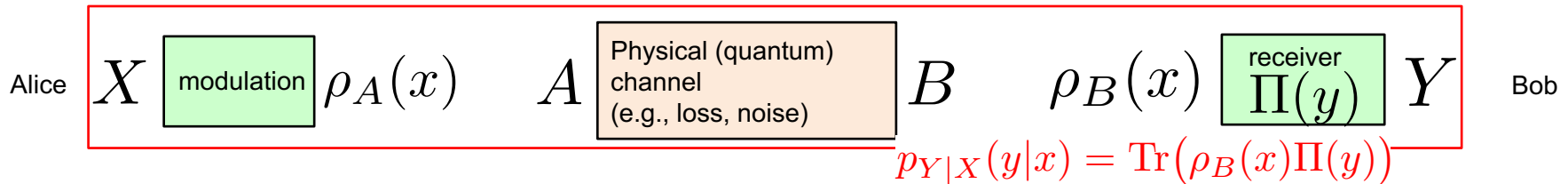
Holevo, 1996
Schumacher,
Westmoreland, 1997



Holevo capacity, $C = \sup_n \frac{\max_{p_{X^n}(\mathbf{x}), \rho_{A^n}(\mathbf{x})} S(\sum p_{X^n}(\mathbf{x}) \rho_{B^n}(\mathbf{x})) - \sum p_{X^n}(\mathbf{x}) S(\rho_{B^n}(\mathbf{x}))}{n}$

“Superadditivity” on the transmitter and receiver side

- Superadditive capacity
 - **Transmitter side:** Using entangled states help achieve higher capacity
 - **Receiver side:** Quantum joint detection receiver helps achieve higher capacity
- Channel not transmitter-side superadditive, it can still be receiver-side
 - Product state modulation is optimal (entanglement at transmitter doesn't help)
 - Holevo Capacity, $C = \max_{p_X} S \left(\sum p_X(x) \rho_B(x) \right) - \sum p_X(x) S(\rho_B(x))$



- Shown : receiver uses symbol by symbol detection; Shannon capacity $C_1 < C$
- Joint detection receiver (JDR) required, unless $\rho_B(x)$ orthogonal pure states
- Calculation of C does not need receiver. Finding receiver achieving C is hard

The single-mode bosonic channel

- Mean photon number per mode constraint at transmitter $\langle \hat{a}_S^\dagger \hat{a}_S \rangle \leq N_S$
- Environment mode in thermal state, mean photon number $\langle \hat{a}_B^\dagger \hat{a}_B \rangle = N_B$

$$\rho_B = \int \frac{1}{\pi N_B} e^{-|\beta|^2/N_B} |\beta\rangle \langle \beta| d^2\beta$$

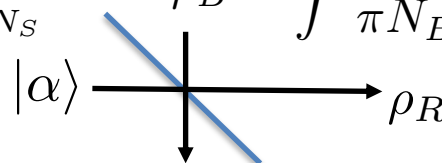
\hat{a}_S \hat{a}_B $\hat{a}_R = \sqrt{\eta} \hat{a}_S + \sqrt{1-\eta} \hat{a}_B$ \hat{a}_E $\eta \in (0, 1]$

- Any linear lossy propagation medium can be decomposed into a collection of single mode bosonic channels

Holevo capacity of the single-mode bosonic channel

$$C = g(\eta N_S + (1 - \eta) N_B) - g((1 - \eta) N_B) \quad \text{bits per mode}$$

- $C >$ Shannon capacity of all known standard optical receivers
- Product coherent state modulation achieves capacity
- Need joint detection receiver (collective measurement over codeword)

$$p(\alpha) = \frac{1}{\pi N_S} e^{-|\alpha|^2 / N_S}$$

$$\rho_B = \int \frac{1}{\pi N_B} e^{-|\beta|^2 / N_B} |\beta\rangle\langle\beta| d^2\beta$$
$$\rho_R = \int \frac{1}{\pi(1 - \eta) N_B} e^{-|\beta - \sqrt{\eta}\alpha|^2 / (1 - \eta) N_B} |\beta\rangle\langle\beta| d^2\beta$$

$\eta \in (0, 1]$

$$g(N) = (1 + N) \log(1 + N) - N \log(N)$$

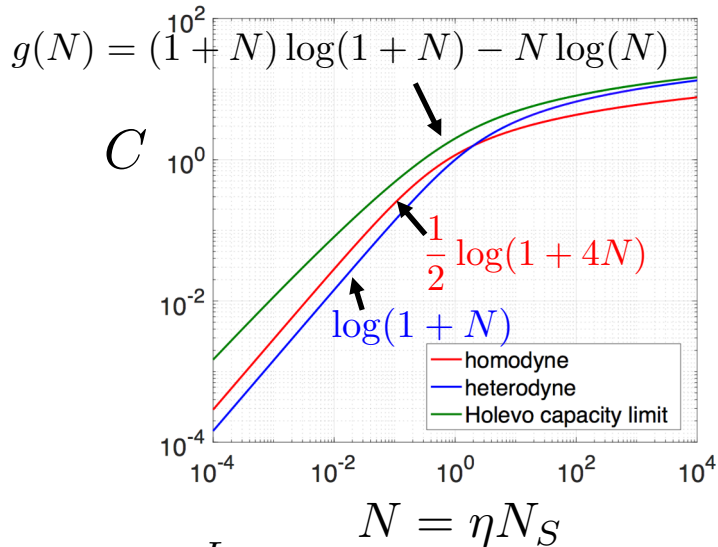
V. Giovannetti, SG, S. Lloyd, L. Maccone, J. H. Shapiro, and H. P. Yuen, *Phys. Rev. Lett.* **92**, 027902 (2004)

V. Giovannetti, SG, S. Lloyd, L. Maccone, and J. H. Shapiro, *Phys. Rev. A* (2004)

V. Giovannetti, R. García-Patrón, N. J. Cerf, and A. S. Holevo, *Nat. Photonics* **8**, 796 (2014)

Quantum limit of classical (optical) communications over a lossy channel, with an input mean photon number constraint

$$N_B = 0$$



$$\eta \sim e^{-\alpha L} \text{ fiber}$$

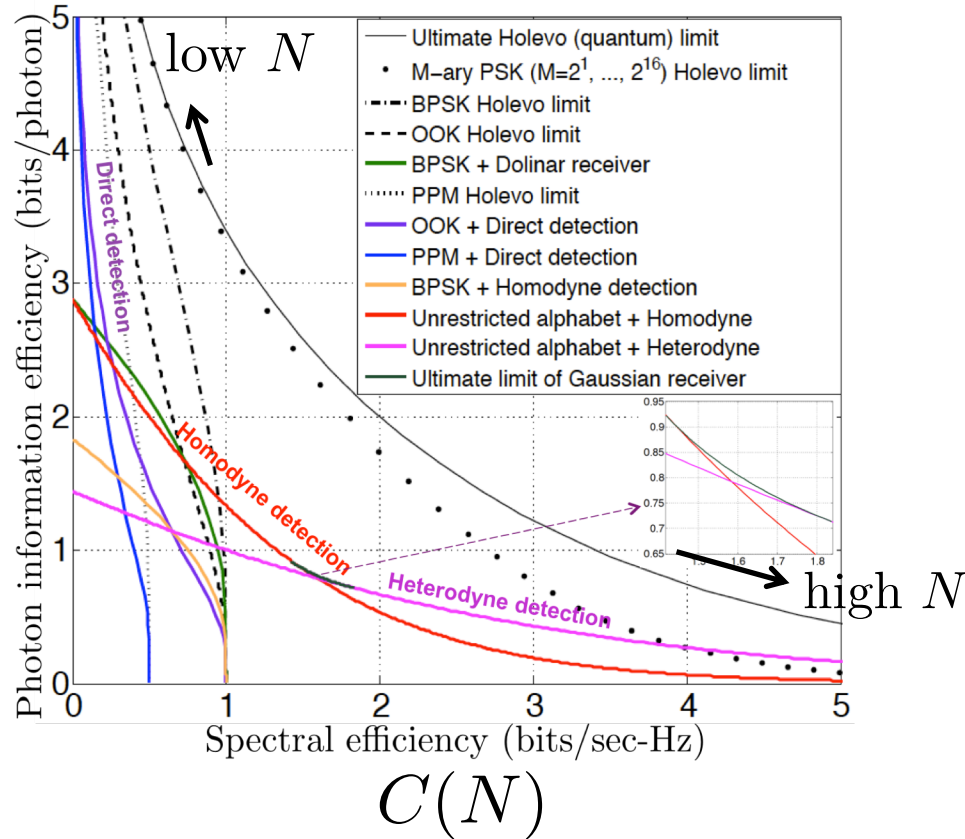
$$\eta \sim 1/L^2 \text{ free-space}$$

η : Transmissivity

Mean photon number

N_S : transmitted per mode

$$\frac{C(N)}{N}$$



Giovannetti, SG, Lloyd, Maccone, Shapiro, & Yuen, PRL (2004)

Takeoka and SG, Physical Review A 89, 042309 (2014)

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Binary hypothesis test: $|\psi_1\rangle$ (hypothesis H_1) vs. $|\psi_2\rangle$ (hypothesis H_2)

Prior probabilities: $p_1 = p$, $p_2 = 1 - p$

Inner product between the two states

$$\begin{aligned}\langle\psi_1|\psi_2\rangle &= \sigma \\ &= \cos 2\theta\end{aligned}$$

$$|1\rangle = \begin{pmatrix} 0 \\ 1 \end{pmatrix}$$

$$|w_1\rangle = \begin{pmatrix} \cos \phi \\ \sin \phi \end{pmatrix}$$

$$|\psi_1\rangle = \begin{pmatrix} \cos \theta \\ \sin \theta \end{pmatrix}$$

$$|\psi_2\rangle = \begin{pmatrix} \cos \theta \\ -\sin \theta \end{pmatrix}$$

$$|w_2\rangle = \begin{pmatrix} \sin \phi \\ -\cos \phi \end{pmatrix}$$

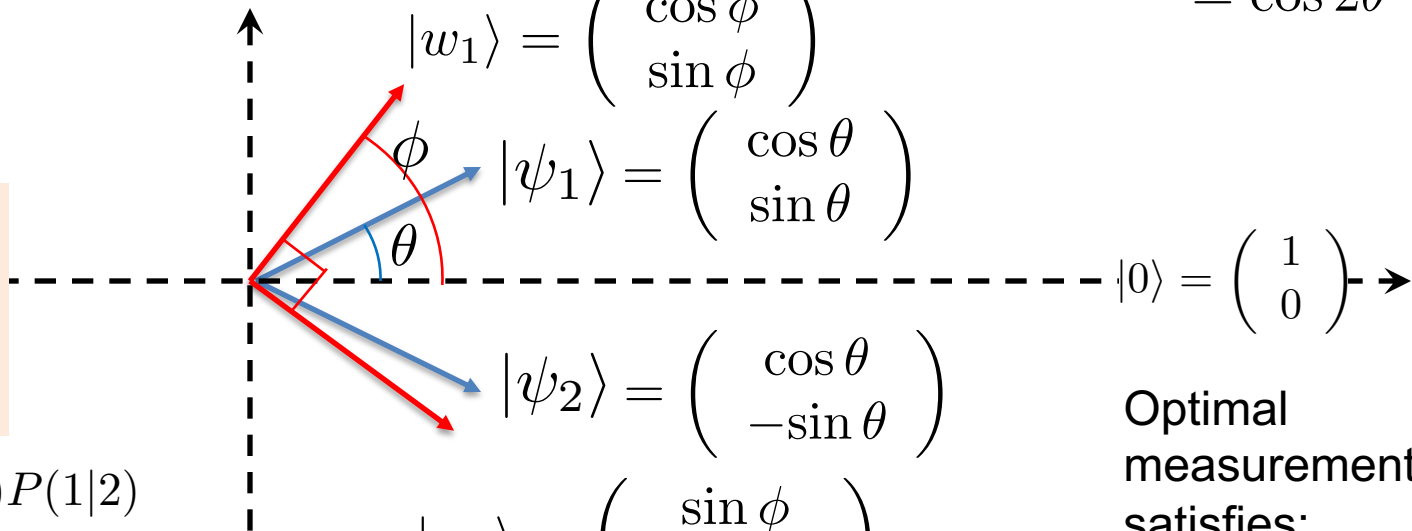
$$|0\rangle = \begin{pmatrix} 1 \\ 0 \end{pmatrix}$$

Optimal measurement satisfies:
 $\tan 2\phi = \frac{\tan 2\theta}{2p - 1}$

Helstrom bound
 (for minimum probability of error state discrimination)

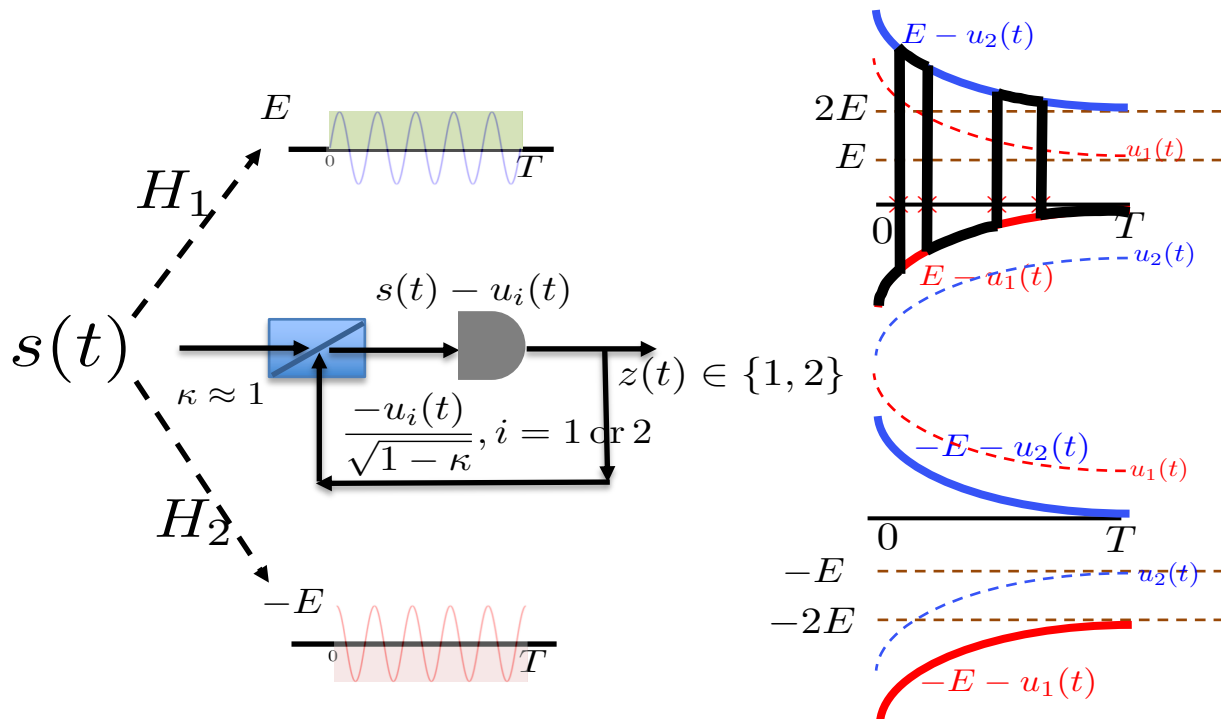
$$P_e = pP(2|1) + (1 - p)P(1|2)$$

$$P_{e,\min} = \frac{1}{2} \left[1 - \sqrt{1 - 4p(1 - p)\sigma^2} \right]$$



Binary phase shift keying (BPSK) modulation

- Consider BPSK modulation, $\{|\alpha\rangle, |-\alpha\rangle\}$, priors $(p, 1 - p)$, $N = |\alpha|^2$



$$\langle \alpha | -\alpha \rangle = e^{-2N}$$

Minimum error probability:

$$P_{e,\min} = \frac{1}{2} \left[1 - \sqrt{1 - 4p(1 - p)e^{-4N}} \right]$$

Example of receiver-side superadditive capacity

- Maximum capacity with the optimal symbol-by-symbol measurement

$$C_1 = 1 - h_2(p), \quad p = \frac{1}{2}[1 - \sqrt{1 - e^{-4N}}] \quad \text{bits per mode}$$

- Holevo capacity (joint detection over infinite-length modulated block)

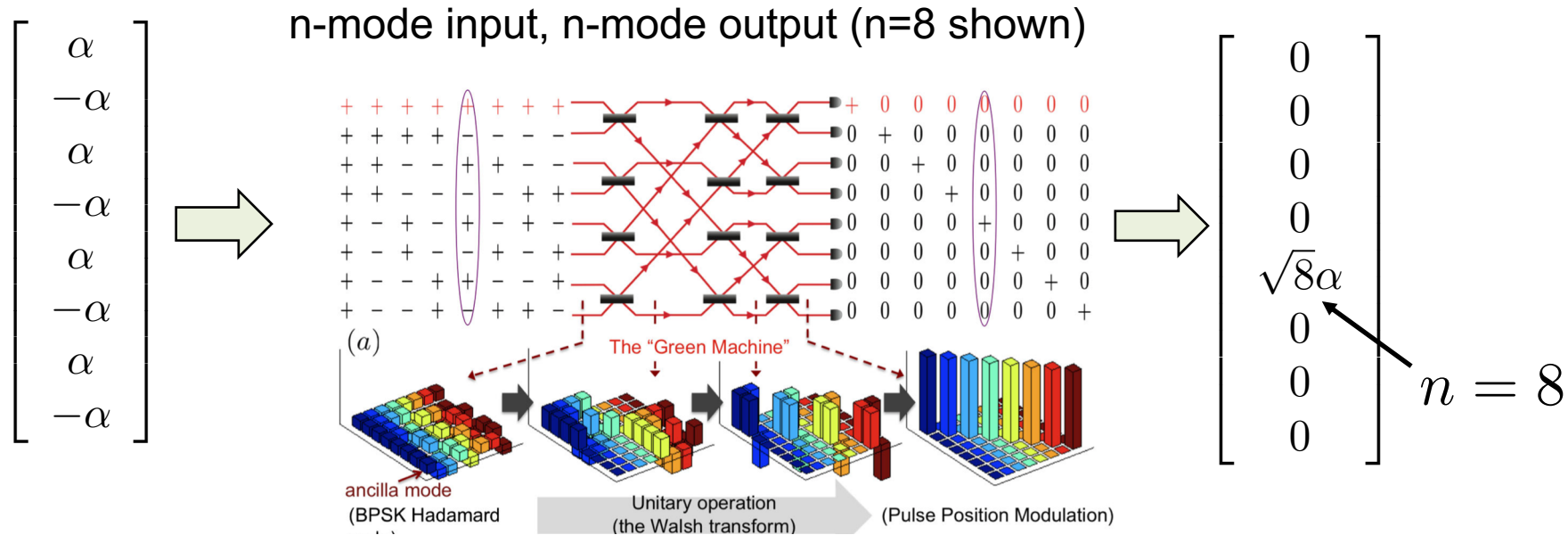
$$C_\infty = S \left(\frac{|\alpha\rangle\langle\alpha| + |-\alpha\rangle\langle-\alpha|}{2} \right) = h_2 \left(\frac{1 + e^{-2N}}{2} \right) \quad \text{bits per mode}$$

- The joint-detection gain most pronounced at low photon number

$$\lim_{N \rightarrow 0} \frac{C_\infty}{C_1} = \infty$$

We do not know, $C_2, C_3, \dots, C_n, \dots$

Joint detection receiver (JDR) for superadditive capacity



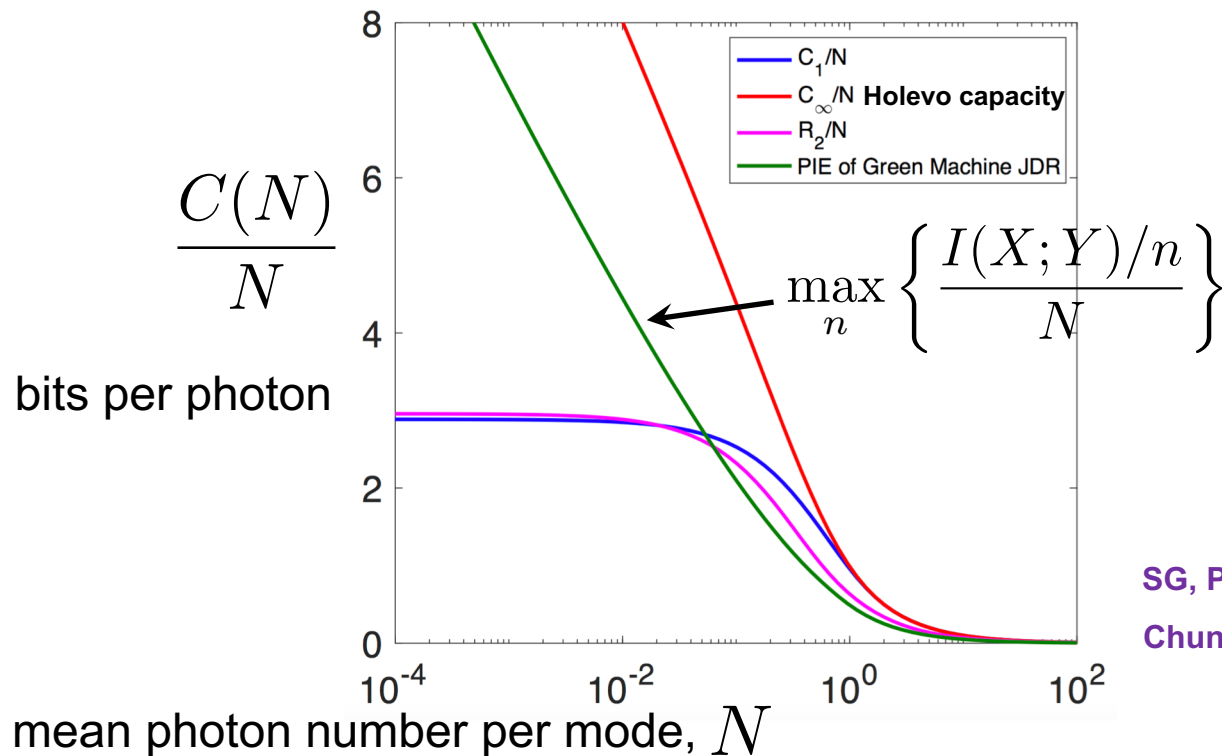
Mean photon number in the n-mode Hadamard codeword (input and output), $n|\alpha|^2 = nN$

$$X \in \{1, 2, \dots, n\} \rightarrow p_{Y|X}(y|x) = \begin{cases} 1 - e^{-nN}, & \text{if } y = x \\ 0, & \text{if } y \neq x \\ e^{-nN}, & \text{if } y = n + 1 \end{cases} \rightarrow Y \in \{1, 2, \dots, n, n + 1\}$$

Superadditive capacity

Bridging the remaining gap to the Holevo capacity requires JDRs that use **quantum optical effects**

- Plotting photon information efficiency, $C(N)/N$



SG, Phys. Rev. Lett. 106, 240502 (2011)

Chung, SG, Zheng, PRA 96, 012320 (2017)

A receiver (based on quantum “belief propagation” to achieve the Holevo capacity

Coherent state binary-phase shift keyed (BPSK) codebook, with eight 5-mode codewords

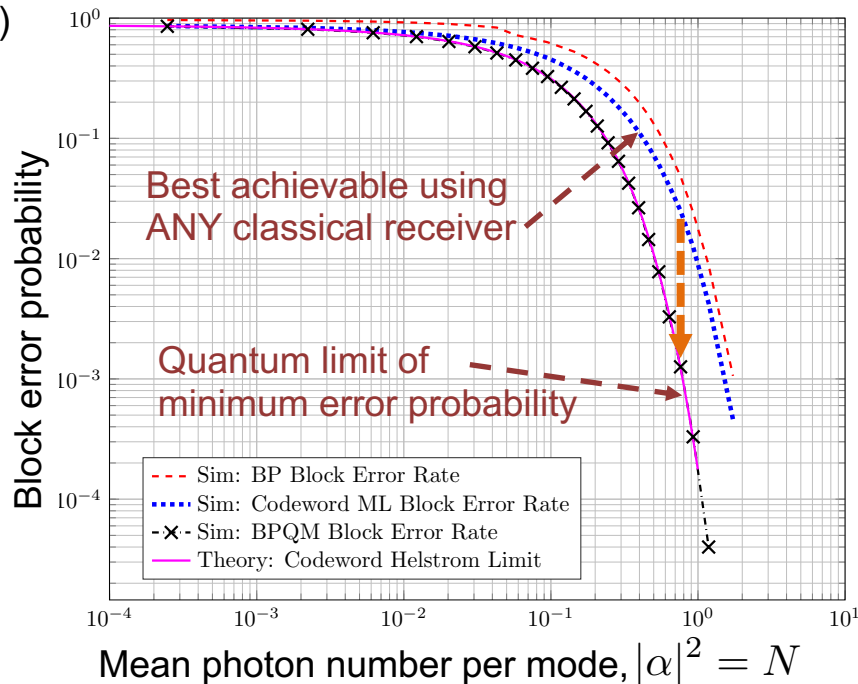
$$\left\{ \begin{array}{l} |-\alpha\rangle|-\alpha\rangle|\alpha\rangle|-\alpha\rangle|\alpha\rangle \\ |\alpha\rangle|\alpha\rangle|\alpha\rangle|\alpha\rangle|\alpha\rangle \\ |-\alpha\rangle|-\alpha\rangle|\alpha\rangle|\alpha\rangle|-\alpha\rangle \\ |\alpha\rangle|\alpha\rangle|\alpha\rangle|-\alpha\rangle|-\alpha\rangle \\ |\alpha\rangle|-\alpha\rangle|-\alpha\rangle|\alpha\rangle|\alpha\rangle \\ |\alpha\rangle|-\alpha\rangle|-\alpha\rangle|-\alpha\rangle|-\alpha\rangle \\ |-\alpha\rangle|\alpha\rangle|-\alpha\rangle|\alpha\rangle|-\alpha\rangle \\ |-\alpha\rangle|\alpha\rangle|-\alpha\rangle|-\alpha\rangle|\alpha\rangle \end{array} \right\}$$

$$|\psi\rangle \longrightarrow U \longrightarrow |\psi_{\pm}\rangle = \sqrt{\frac{1+x}{2}}|0\rangle \pm \sqrt{\frac{1-x}{2}}|1\rangle$$

$$|\psi\rangle \in \{|\alpha\rangle, |-\alpha\rangle\}$$

$$\langle\alpha|-\alpha\rangle = e^{-2N} \equiv x$$

$$\langle\psi_+|\psi_-\rangle = x$$



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Entanglement assisted classical communications

- Communication capacity, when Alice and Bob pre-share (unlimited amount of) entanglement, $C_E = \max_{\rho_A \in \mathcal{H}_A} I(\mathcal{N}^{A \rightarrow B}, \rho_A)$

$$I(\mathcal{N}^{A \rightarrow B}, \rho_A) = S(\rho_A) + S(\mathcal{N}^{A \rightarrow B}(\rho_A)) - S((\mathcal{N}^{A \rightarrow B} \otimes \mathbb{I})[\Phi])$$

– where, Φ is a purification of ρ_A

Entanglement assisted capacity of the bosonic channel

- Mean photon number constraint, N_S

- Take, $\rho_A = \int \frac{1}{\pi N_S} e^{-|\alpha|^2} |\alpha\rangle\langle\alpha| d^2\alpha$

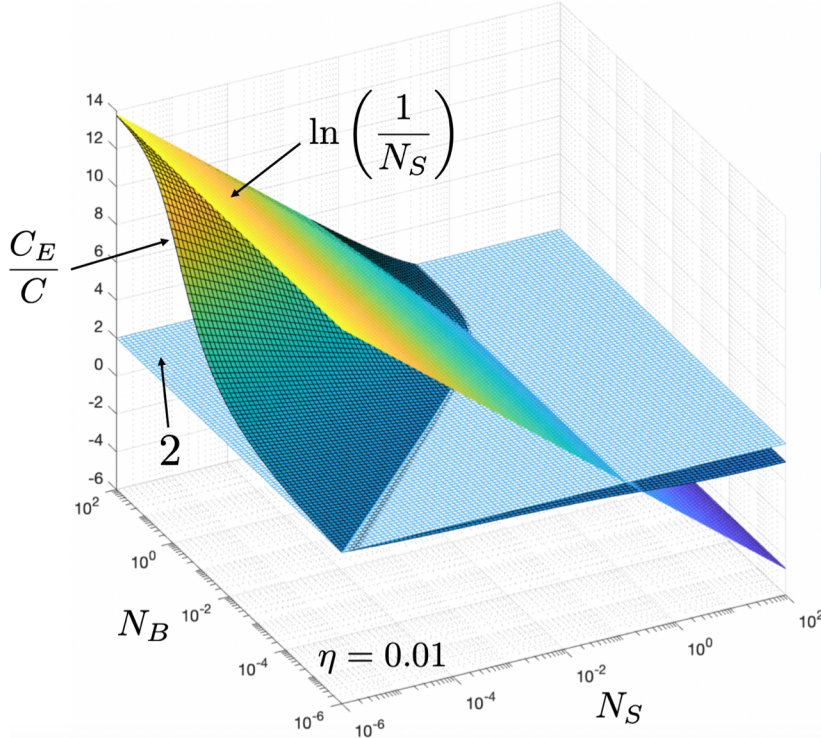
- Purification, $|\Phi\rangle = \sum_{n=0}^{\infty} \sqrt{\frac{N_S^n}{(1+N_S)^{1+n}}} |n\rangle_S |n\rangle_I$

- Capacity, $C_E(\eta, N_S, N_B) = g(N_S) + g(N'_S) - g(A_+) - g(A_-)$

- $A_{\pm} = (D - 1 \pm (N'_S - N_S))/2$

- $D = \sqrt{(N_S + N'_S + 1)^2 - 4\eta N_S(N_S + 1)}$

Comparison of C_E with C



The entanglement assistance is most pronounced, i.e., $C_E/C \sim \ln(1/N_S)$ when $N_S \ll 1, N_B \gg 1$

Low brightness transmitter
 Long center wavelength (high noise)

Taylor series expansion at $N_S = 0$ yields:

$$C_E = -N_S \ln N_S + o(N_S)$$

$$C = N_S \ln \left(1 + [(1 - \eta)N_B]^{-1} \right) + o(N_S)$$

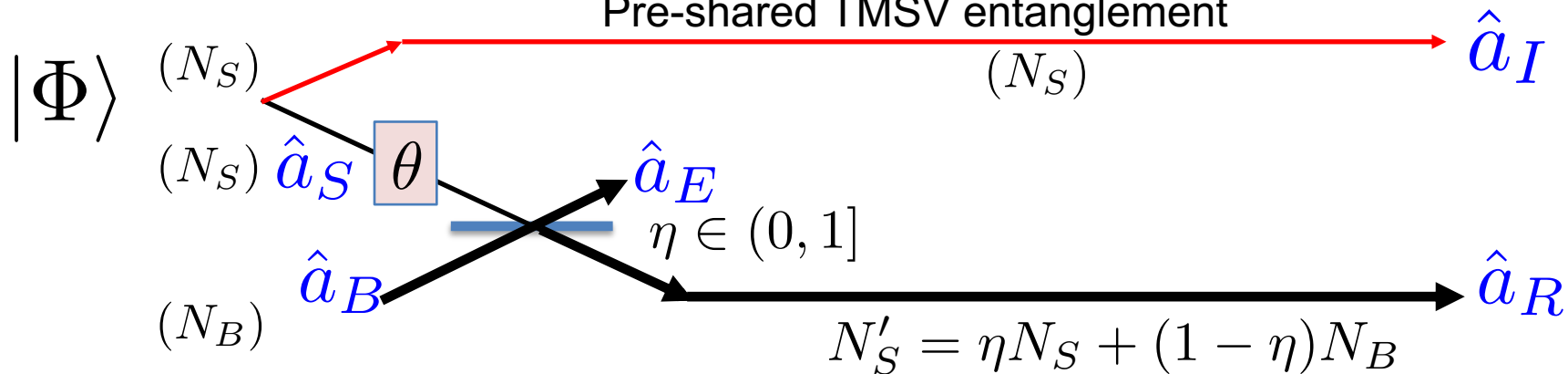
$$\lim_{N_S \rightarrow 0} \frac{C_E}{C \ln(1/N_S)} = \frac{1}{(1 + (1 - \eta)N_B) \ln \left(1 + [(1 - \eta)N_B]^{-1} \right)}$$

Achieving C_E for the bosonic channel

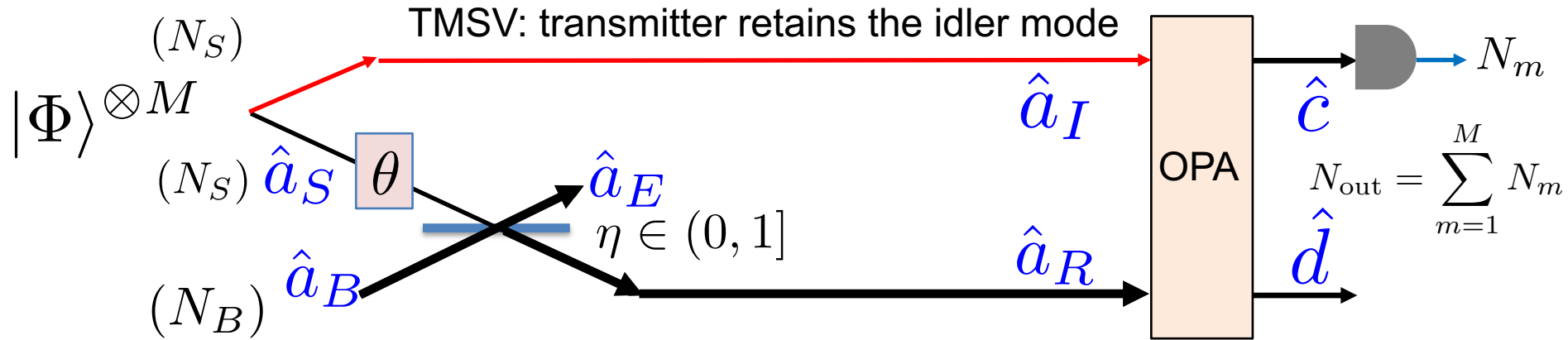
- A two-mode squeezed vacuum (TMSV) source, and phase modulation achieves the entanglement assisted capacity C_E , when $N_S \ll 1, N_B \gg 1$
 - In general, active modulation (squeezing, displacement) of TMSV maybe necessary
 - binary phase modulation suffices

$$|\Phi\rangle = \sum_{n=0}^{\infty} \sqrt{\frac{N_S^n}{(1+N_S)^{1+n}}} |n\rangle_S |n\rangle_I$$

Pre-shared TMSV entanglement



The Optical Parametric Amplifier (OPA) based receiver



$$\langle \hat{a}_R \hat{a}_I \rangle = \sqrt{\eta N_S (N_S + 1)}$$

$$\frac{C_E}{C} \rightarrow 2$$

$$\hat{c}^{(m)} = \sqrt{G} \hat{a}_I^{(m)} + \sqrt{G-1} \hat{a}_R^{\dagger(m)}$$

$$\hat{d}^{(m)} = \sqrt{G} \hat{a}_R^{(m)} + \sqrt{G-1} \hat{a}_I^{\dagger(m)}$$

$$E[N_m] = \langle \hat{c}^\dagger \hat{c} \rangle = GN_S + (G-1)(1 + (1-\eta)N_B + \eta N_S) \pm 2\sqrt{G(G-1)}\sqrt{\eta N_S(N_S+1)}$$

Receiver based on sum frequency generation (SFG)

$$\hat{H}_I = \hbar g \sum_{m=1}^M \left(\hat{b}^\dagger \hat{a}_{S_m} \hat{a}_{I_m} + \hat{b} \hat{a}_{S_m}^\dagger \hat{a}_{I_m}^\dagger \right) \text{ SFG}$$

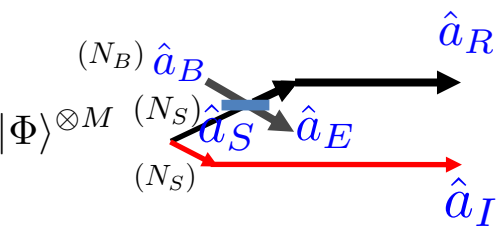
The SF (\hat{b}_k) modes are in displaced thermal states:

$$\hat{\rho}_{\text{th}}(\alpha^{(k)}, N_T) = \int_{\mathbb{C}} \frac{1}{\pi N_T} e^{-|\beta - \alpha^{(k)}|^2 / N_T} |\beta\rangle \langle \beta| d^2\beta$$

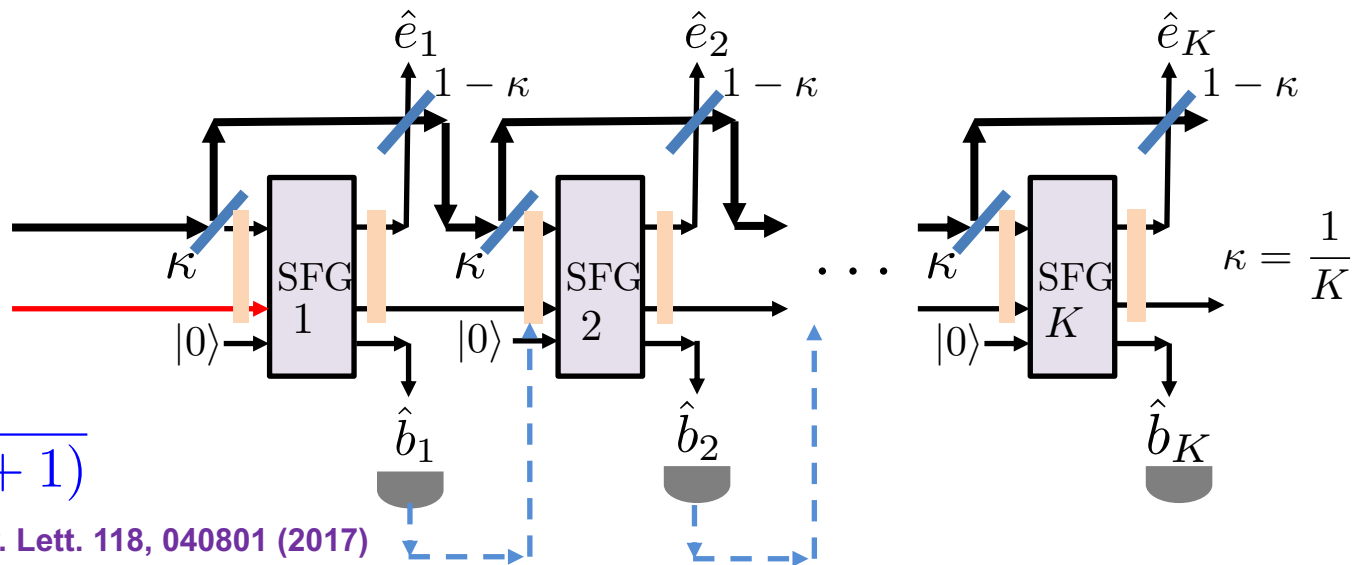
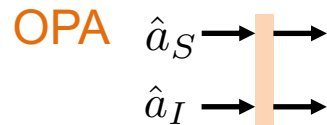
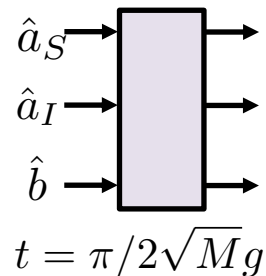
$$\alpha^{(k)} = \sqrt{M\kappa\eta N_S(1 + N_S)} \mu^{k-1}$$

$$\mu = (1 - \kappa(1 + N'_S))^2$$

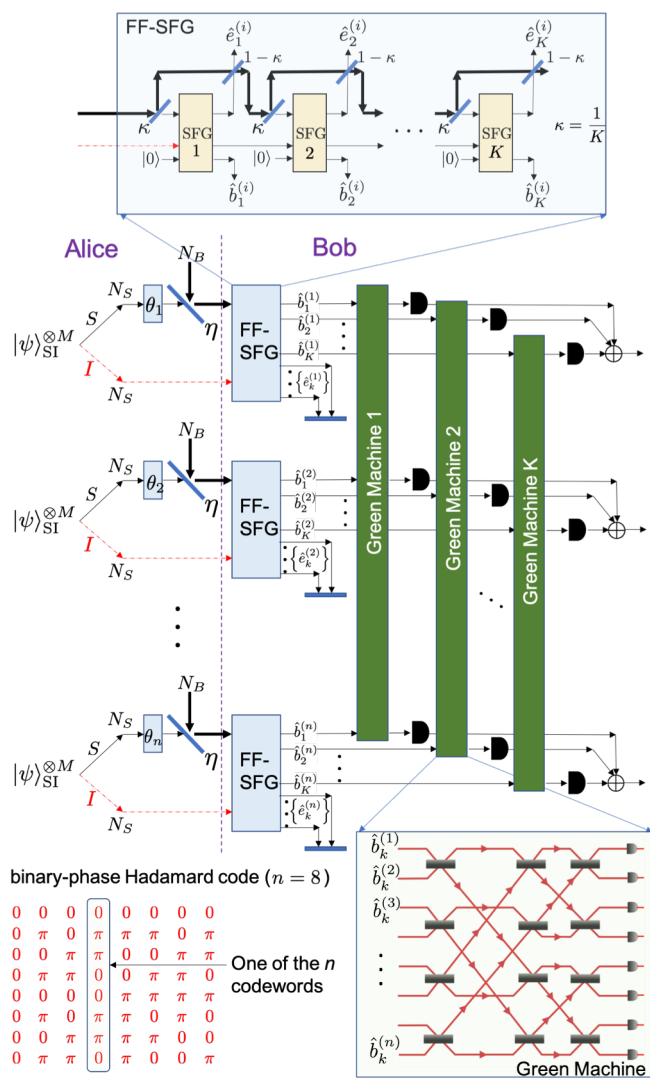
$$N_T = \kappa N_S N'_S$$



$$\langle \hat{a}_R \hat{a}_I \rangle = \sqrt{\eta N_S (N_S + 1)}$$



Transmitter and joint-detection receiver for entanglement assisted communication



k-th Green Machine's inputs:

$$\hat{\rho}_{\text{th}}(\pm\alpha^{(k)}, N_T)$$

k-th Green Machine's outputs (one of them has this mean, others have zero mean)

$$\hat{\rho}_{\text{th}}(\sqrt{n}\alpha^{(k)}, N_T)$$

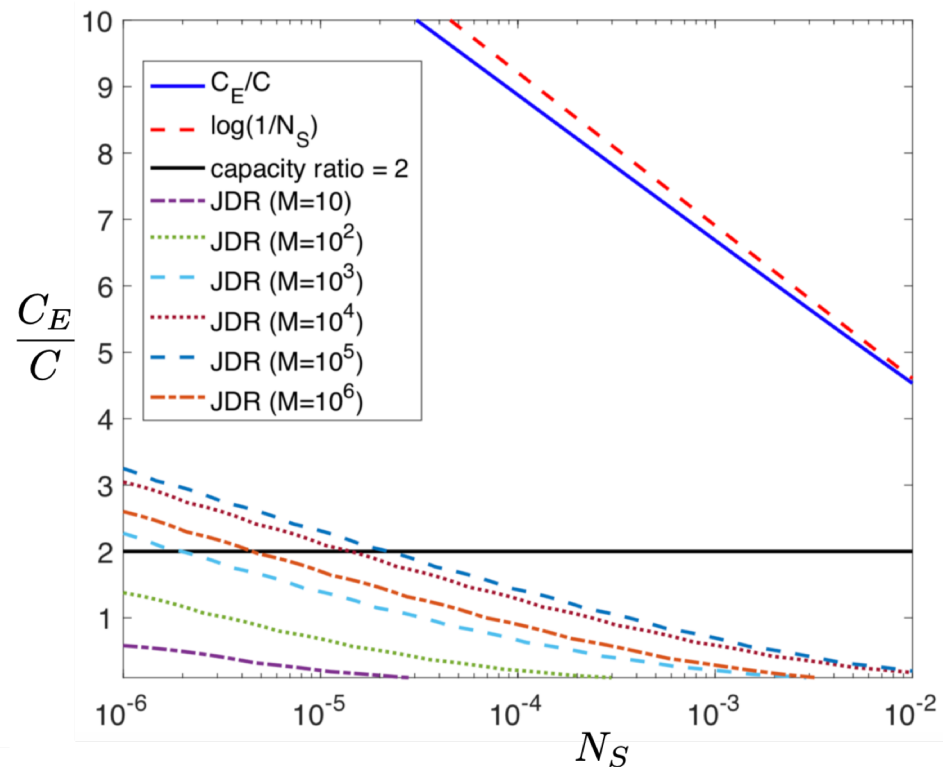
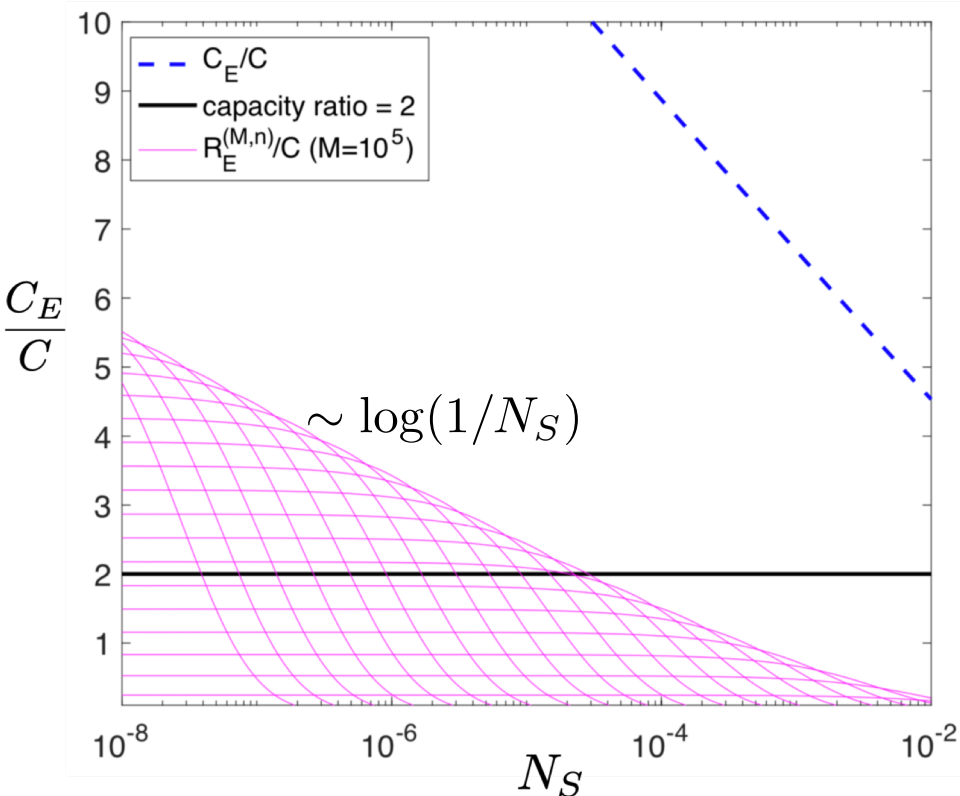
$$\langle 0 | \hat{\rho}_{\text{th}}(\alpha, N_T) | 0 \rangle = (1/(N_T + 1))e^{-|\alpha|^2/(N_T + 1)}$$

$$\alpha^{(k)} = \sqrt{M\kappa\eta N_S(1 + N_S)}\mu^{k-1}$$

$$\mu = (1 - \kappa(1 + N'_S))^2$$

$$N_T = \kappa N_S N'_S$$

Superadditivity in entanglement assisted communication rate



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Covert communications

- Provably covert communications needs $N_S = O(1/\sqrt{n})$, resulting in:

$$O(\sqrt{n}) \text{ bits in } n \text{ modes}$$

B. A. Bash, A. H. Gheorghe, M. Patel, J. L. Habif, D. Goeckel, D. Towsley, and SG, *Nat. Commun.* 6, 8626 (2015)

- Entanglement assisted covert communications: the $\log(1/N_S)$ enhancement in capacity results in:

$$O(\sqrt{n} \log n) \text{ bits in } n \text{ modes}$$

Gagatsos, Bullock, Bash, [arXiv:2002.06733 \[quant-ph\]](https://arxiv.org/abs/2002.06733) (2020)

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Conclusions and ongoing work

- **Holevo capacity, C** : Quantum limit of classical communications
 - (product) coherent state, i.e., laser light modulation is capacity optimal
 - Attaining ultimate limit of classical communications needs quantum (joint detection) receivers
- **Entanglement assisted capacity, C_E** : highest gain when the transmitter is low brightness and thermal noise is high
 - TMSV pre-shared entanglement and SFG receiver front end translates the problem back to receiver design for coherent states to achieve Holevo capacity
 - Covert communications: shared entangled breaks the “square root law”
- Find better transmitter modulation, codes, and receivers to fully achieve C_E : explore **resource theories** of non-Gaussian components
- Connection with **quantum illumination radar**

Thank you!

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